

wed 7 Dec 2016

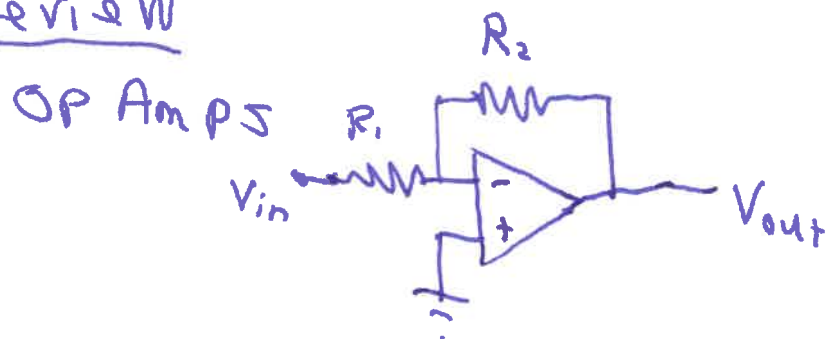
143

Final Friday 16 Dec, 221 HA
3:30 - 5:30

Check Blackboard (Update tonight?)

Watch Website for Office Hours + more
Email Re: Final Conflicts

Review



No current into op-Amp inputs

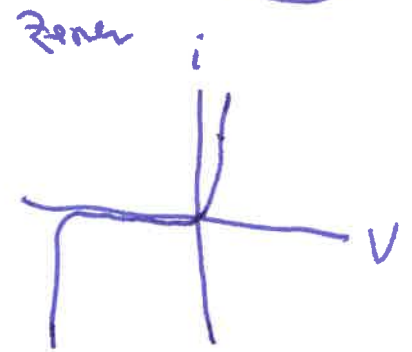
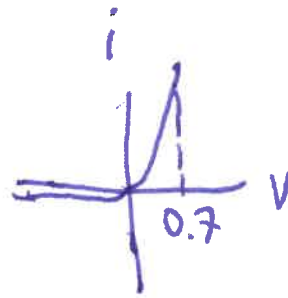
$$V_+ - V_- = 0 \text{ if possible}$$

"Virtual Ground"

Gain $A_v = \frac{V_{out}}{V_{in}}$

Input + Output impedances.

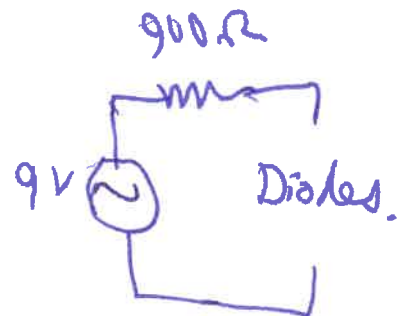
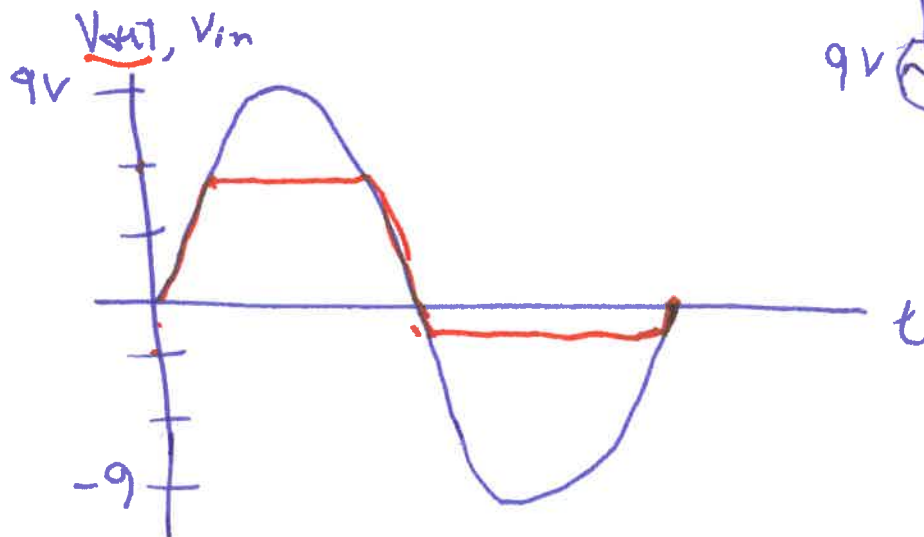
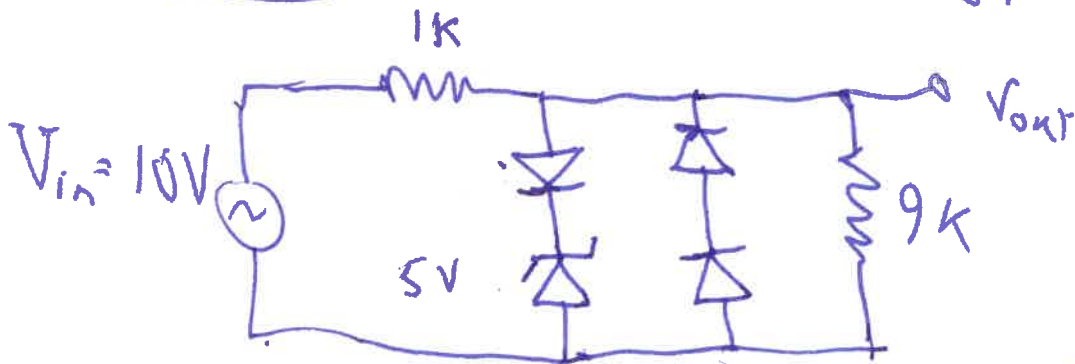
Diodes

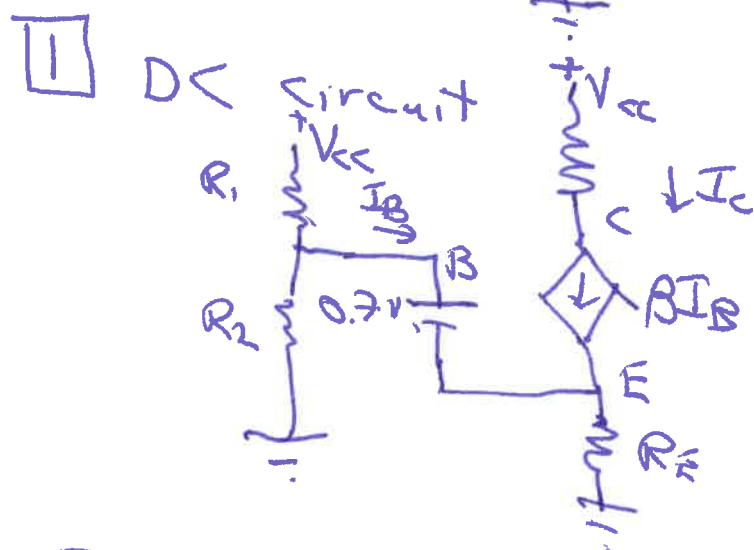
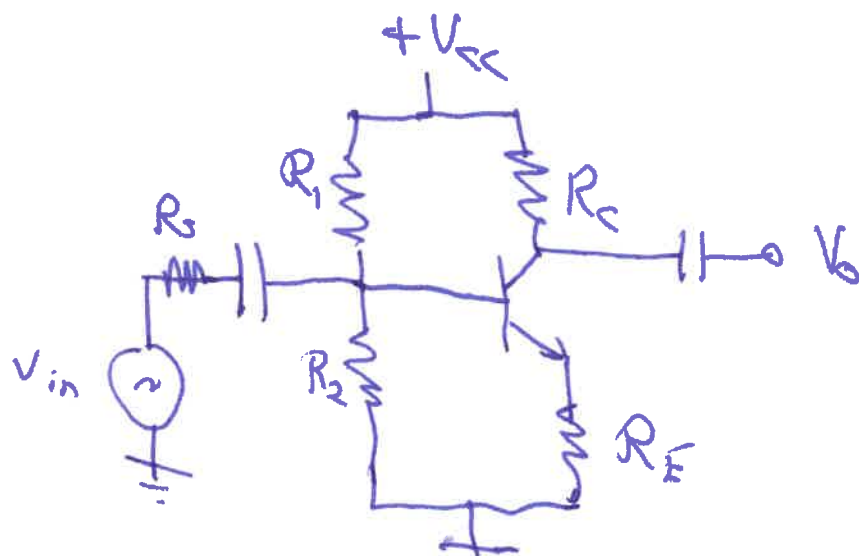


Shockley Eq.

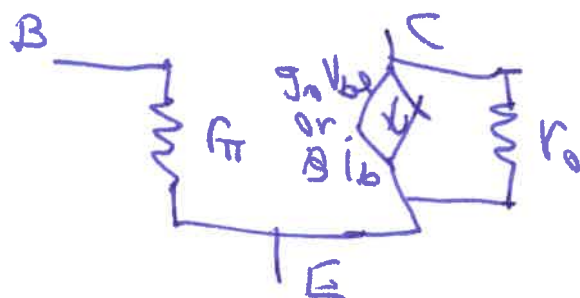
$$i = I_s (-1 + e^{V/nV_T})$$

Example





2] Small signal model of the transistor



$$r_{\pi} = \frac{\beta V_T}{I_C}$$

$$g_m = \frac{I_C}{V_T}$$

3] AC circuit (whole circuit)

$$\text{Gain: } A_v = \frac{V_{out}}{V_{in}}$$

Impedances.

BJT amplifiers.

CE high gain

CC $A_v \approx 1$ CB $A_i = 1$ FET~~Triode~~ ~~Saturation~~

Saturation

$$i_D = K \left[2(V_{GS} - V_{TO})V_{DS} - V_{DS}^2 \right]$$

$$i_D = K(V_{GS} - V_{TO})^2$$

Boundary

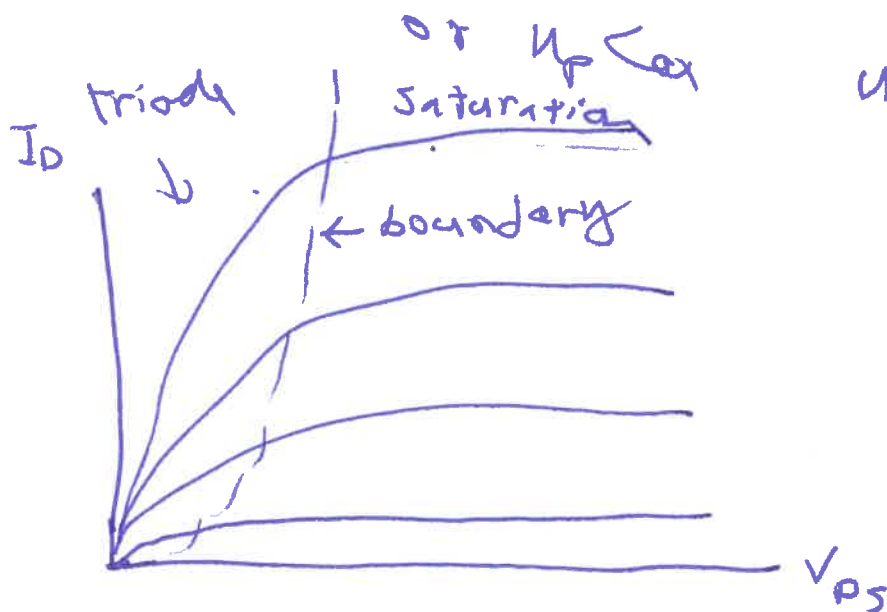
$$I_D = K V_{DS}^2 \quad \text{or} \quad V_{DS} = V_{GS} - V_{TO}$$

$$K = \frac{W}{L} \frac{K_P}{2}$$

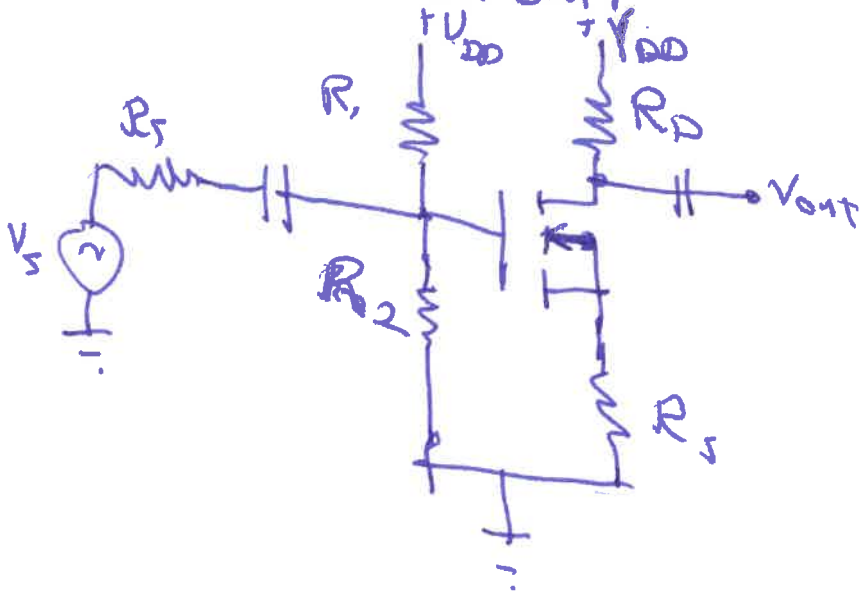
$$K_P = \mu_n C_{ox} \approx 50 \mu A/V^2$$

$$\text{or } \mu_p C_{ox}$$

$$\mu_p \approx \frac{\mu_n}{2}$$



I DC Circuit

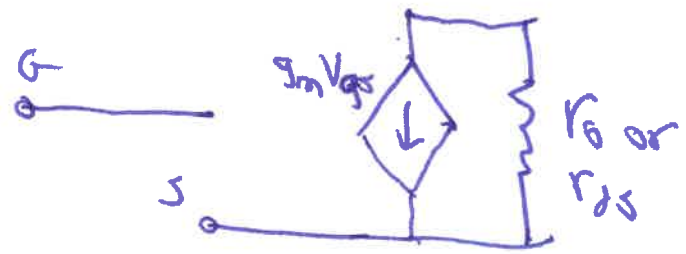


Quadratic E_n .

$$I_G = 0$$

$$I_S = I_D$$

2 Small Signal Model



$$g_m = 2k(V_{GS} - V_{th})$$

$$= 2I_D / (V_{GS} - V_{th})$$

$$= 2\sqrt{kI_D}$$

$$r_o = \frac{V_A}{I_D}$$

AC circuit

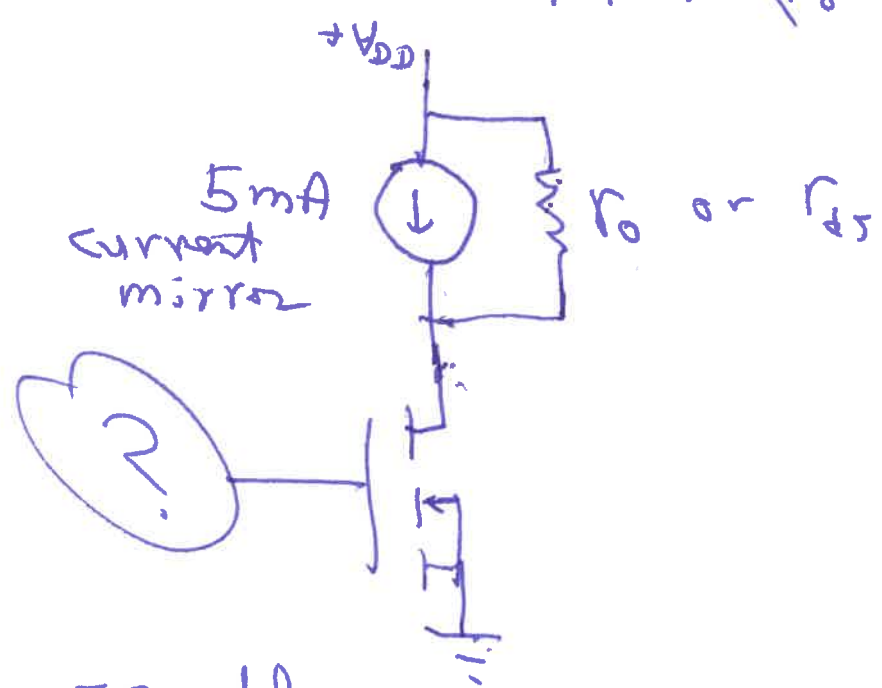
Gain $A_v = \frac{V_{out}}{V_{in}}$

Impedances.

- C_S - High Gain
- C_D - $A_r \approx 1$
- C_G - $A_i = 1$

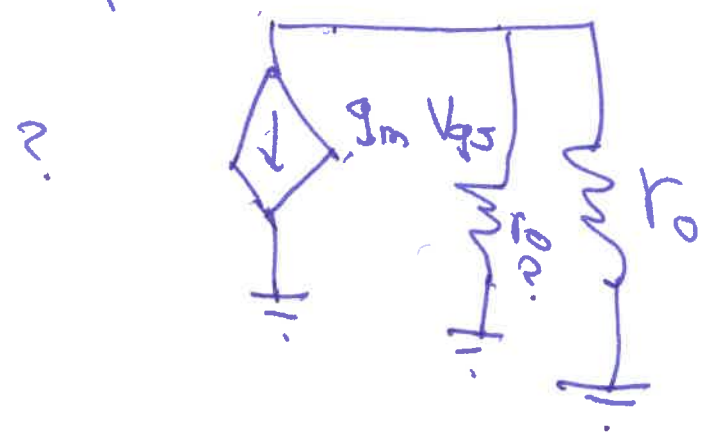
Current mirrors + Active load

FET with (r_o or r_{ds})

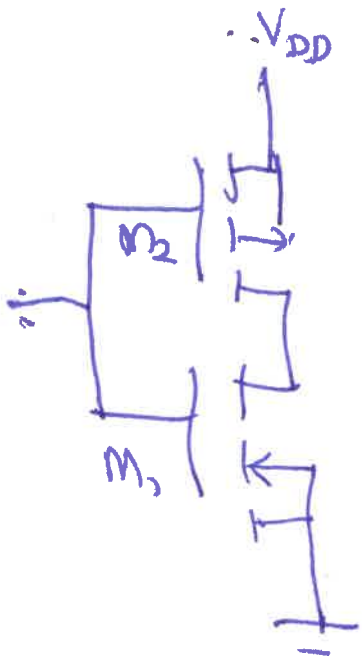


$$r_o = \frac{100V}{5mA} = 20K$$

SS model

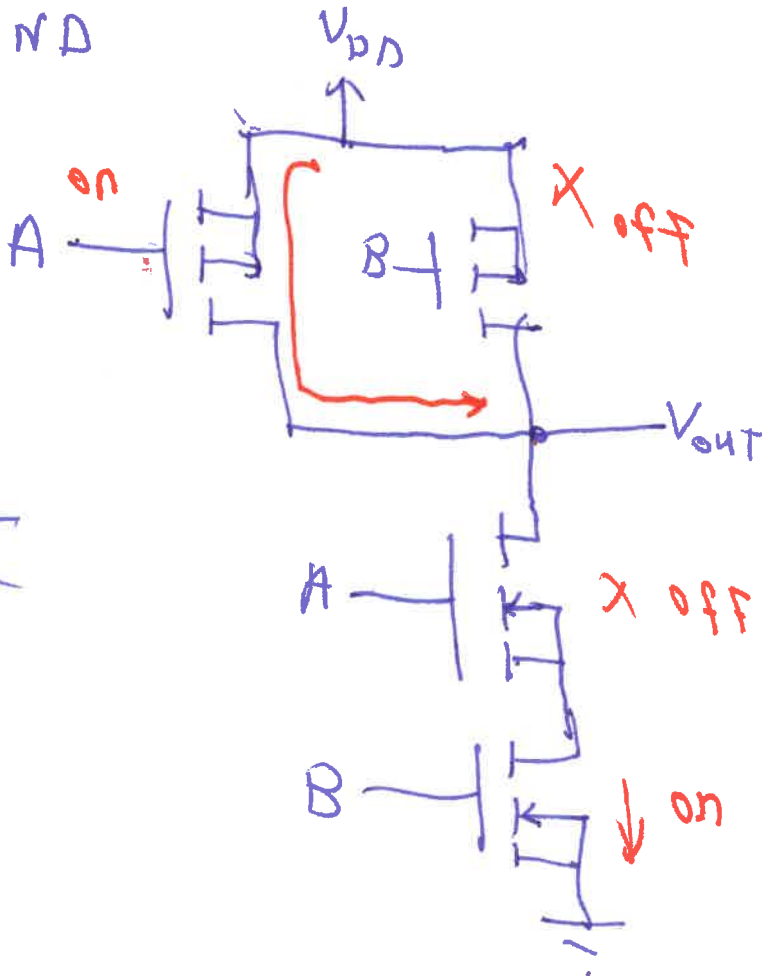


CMOS logic



Inverter

NAND



A	B	V _{out}
L	L	H
L	H	L
H	L	L
H	H	H

